

16 Bit, 400KSPS, Analog-to-Digital Converter

1 Features

- **16 Bits No Missing Codes**
- **Very Low Noise: 45 μ Vrms**
- **Excellent Linearity:**
 - ± 2 LSB typ INL
 - ± 0.5 LSB typ DNL
 - ± 4 LSB typ Offset
 - ± 6 LSB typ Gain Error
- **microPower:**
 - 13mW at 5V, 400KSPS
 - 4.86mW at 2.7V, 300KSPS
 - 1.62mW at 2.7V, 100KSPS
 - 162 μ W at 2.7V, 10KSPS
- **MSOP8 Packages**
- **SPI Interface**

2 Applications

- **Automotive Navigation**
- **FA or ATM Equipment**
- **Industrial Controls**
- **Robotics**
- **Battery-Operated Systems**
- **Instrumentation and Control Systems**

3 Descriptions

The ZMB1430B is a 16-bit, sampling, analog-to-digital (A/D) converter specified for a supply voltage range from 2.7V to 5.5V. It requires very little power, even when operating at the full data rate. At lower data rates, the high speed of the device enables it to spend most of its time in the power-down mode. For example, the average power dissipation is less than 162 μ W at a 10kHz data rate.

The ZMB1430B offers excellent linearity and very low noise and distortion. It also features a synchronous serial (SPI/SSI-compatible) interface and a pseudo-differential input. The reference voltage can be set to any level within the range of 0.1V to VDD.

Low power and small size make the ZMB1430B ideal for portable and battery-operated systems. It is also a perfect fit for remote data-acquisition modules, simultaneous multichannel systems, and isolated data acquisition. The ZMB1430B is available in an MSOP8 package.

Device Information ⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
ZMB1430B	MSOP8	3.0mm×4.9mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

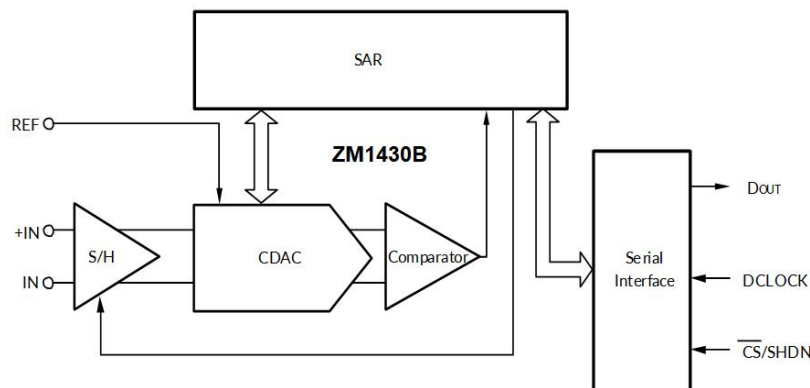


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4 Revision History

Note: Page numbers for previous revisions may differ from page numbers in the current version.

VERSION	Change Date	Change Item
A.1	2024/03/01	Preliminary version completed

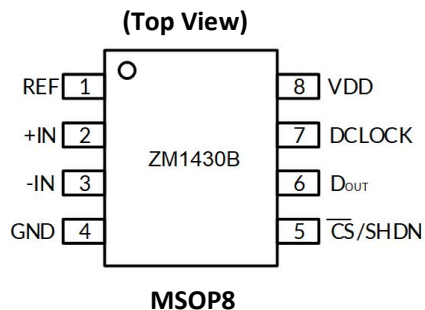
5 Package/Ordering Information ⁽¹⁾

Orderable Device	Package Type	Op Temp(°C)	Device Marking ⁽²⁾	MSL ⁽³⁾	Package Qty
ZMB1430BXM	MSOP8	-40°C ~+140°C	ZMB1430B	MSL1	Tape and Reel,4000

NOTE:

- (1) This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the right-hand navigation.
- (2) There may be additional marking, which relates to the lot trace code information (data code and vendor code), the logo or the environmental category on the device.
- (3) MSL, The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications.

6 Pin Configuration and Functions (Top View)



Pin Description

NAME	PIN	I/O	DESCRIPTION
REF	1	Analog input	Reference input. It must be thoroughly bypassed.
+IN	2	Analog input	Noninverting analog input.
-IN	3	Analog input	Inverting analog input.
GND	4	Power-supply connection	The ground return for the supply.
$\overline{\text{CS}}/\text{SHDN}$	5	Digital output	Chip select when low; Shutdown mode when high.
D _{OUT}	6	Digital output	Digital data output. The output words are clocked out of this pin by the DCLOCK pin.
DCLOCK	7	Digital input	Data clock synchronizes the serial data transfer and determines conversion speed.
VDD	8	Power-supply connection	Power supply. These pins must be connected to a quiet 2.7V to 5.25V source and bypassed to GND with 0.1μF and 1μF monolithic capacitors placed within 1 cm of the power pin.

7 Specifications

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted) ⁽¹⁾

			MIN	MAX	UNIT
Supply voltage, V_{DD} ⁽²⁾			-0.3	6.5	V
Voltage on any analog pin to GND ⁽²⁾			-0.3	$V_{DD}+0.3$	V
Voltage on any digital pin to GND ⁽²⁾			-0.3	$V_{DD}+0.3$	V
Voltage on REF pin to GND ⁽²⁾			-0.3	$V_{DD}+0.3$	V
Input current at any pin (except power supply pins)				± 10	mA
θ_{JA}	Package thermal impedance ⁽³⁾	MSOP8		206	°C/W
Soldering temperature, infrared (10 sec)				215	°C
Operating temperature, T_A			-40	125	°C
Storage temperature, T_{stg}			-65	150	°C

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

(2) Analog input terminal is diode-clamped to the power-supply rails. Input signal that can swing more than 0.3V beyond the supply rails should be current-limited to 10mA or less.

(3) The package thermal impedance is calculated in accordance with JESD-51.

7.2 ESD Ratings

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	± 2000	V
		Charged-device model (CDM), per ANSI/ESDA/JEDEC JS-002 ⁽²⁾	± 1500	

(1) JEDEC document JEP155 states that 500 V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250 V CDM allows safe manufacturing with a standard ESD control process.



ESD SENSITIVITY CAUTION

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

7.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_{DD}	Supply voltage	2.7		5.5	V
REF	Reference input voltage	0.1		V_{DD}	
+IN/-IN	-IN to GND	-0.3	0	0.5	
	+IN to GND	-0.3		$V_{DD}+0.2$	
	+IN - (-IN)	0		V_{REF}	
T_A	Operating temperature	-40		125	°C

7.4 Electrical Characteristics: VDD = +5V

At -40°C to $+125^{\circ}\text{C}$, $V_{\text{REF}} = +5\text{V}$, $-\text{IN} = \text{GND}$, $F_s = 400\text{kHz}$, and $f_{\text{DCLOCK}} = 25 \times F_s$, typical values are at $T_A = 25^{\circ}\text{C}$, unless otherwise noted.
(1)(2)

PARAMETER		CONDITIONS	MIN	TYP	MAX	UNIT
STATIC CONVERTER CHARACTERISTICS						
NMC	Resolution with no missing codes		16			Bits
INL	Integral non-linearity			± 2		LSB
DNL	Differential non-linearity			± 0.5		LSB
V_{OS}	Offset error			± 4		LSB
TCV_{OS}	Offset error drift			± 0.3		ppm/ $^{\circ}\text{C}$
GE	Gain error			± 6		LSB
TCGE	Gain error drift			± 0.4		ppm/ $^{\circ}\text{C}$
TUE	Total Unadjusted Error			± 5		LSB
	Noise			45		μV_{rms}
				4		LSB_{PP}
DYNAMIC CONVERTER CHARACTERISTICS ⁽³⁾						
ENOB	Effective number of bits	$f_{\text{IN}}=2\text{kHz}$		14.69		Bits
		$f_{\text{IN}}=10\text{kHz}$		14.66		Bits
SINAD	Signal-to-noise plus distortion ratio	$f_{\text{IN}}=2\text{kHz}$		90.2		dB
		$f_{\text{IN}}=10\text{kHz}$		90		dB
SNR	Signal-to-noise ratio	$f_{\text{IN}}=2\text{kHz}$		90.5		dB
		$f_{\text{IN}}=10\text{kHz}$		90.2		dB
THD	Total harmonic distortion	$f_{\text{IN}}=2\text{kHz}$		-99		dB
		$f_{\text{IN}}=10\text{kHz}$		-98		dB
SFDR	Spurious-free dynamic range	$f_{\text{IN}}=2\text{kHz}$		101		dB
		$f_{\text{IN}}=10\text{kHz}$		100		dB
FPBW	60dB SINAD bandwidth	5V supply		700		kHz
PSRR	Power-supply rejection	$4.75\text{V} \leq V_{\text{DD}} \leq 5.25\text{V}$		0.5		LSB/V
ANALOG INPUT CHARACTERISTICS						
FSR	Full-scale range	+IN – (-IN)	0		V_{REF}	V
	Common-mode signal	-IN	-0.3		0.5	V
C_{IN}	Input capacitance	-IN = GND, during sampling		48		pF
I_{IL}	Input leakage current	$\overline{\text{CS}}/\text{SHDN}=V_{\text{DD}}$, SCLK off		500		nA

(1) Data sheet minimum and maximum specification limits are ensured by design, test, or statistical analysis.

(2) Applies for 5.0V nominal supply: $V_{\text{DD}}(\text{min}) = 4.5\text{V}$ and $V_{\text{DD}}(\text{max}) = 5.5\text{V}$.

(3) All ac parameters are tested at -0.2 dBFS.

Electrical Characteristics: VDD = +5V (continued)

At -40°C to $+125^{\circ}\text{C}$, $V_{\text{REF}} = +5\text{V}$, $-\text{IN} = \text{GND}$, $F_{\text{S}} = 400\text{kHz}$, and $f_{\text{DCLOCK}} = 25 \times F_{\text{S}}$, typical values are at $T_{\text{A}} = 25^{\circ}\text{C}$, unless otherwise noted.

REFERENCE INPUT CHARACTERISTICS						
V_{REF}	Reference voltage		0.1		V_{DD}	V
C_{REF}	Reference input capacitance			48		pF
I_{REF}	Reference input current	$F_{\text{S}} = 400\text{KSPS}$		200		μA
		$F_{\text{S}} = 300\text{KSPS}$		150		μA
		$F_{\text{S}} = 250\text{KSPS}$		120		μA
		$F_{\text{S}} = 100\text{KSPS}$		50		μA
		$F_{\text{S}} = 50\text{KSPS}$		25		μA
		$F_{\text{S}} = 10\text{KSPS}$		7		μA
		$\overline{\text{CS}}/\text{SHDN} = V_{\text{DD}}$		0.1		μA
SAMPLING DYNAMIC CHARACTERISTICS						
t_{CONV}	Conversion time	$25\text{kHz} \leq f_{\text{DCLOCK}} \leq 10\text{MHz}$	18			T_{DCLOCK}
t_{AQ}	Acquisition time		4.5	5		T_{DCLOCK}
F_{S}	Throughput rate	25 DCLOCKS			400	KSPS
f_{DCLOCK}	Clock frequency		0.025		10	MHz
POWER SUPPLY CHARACTERISTICS						
I_{VDD}	Operating supply current	$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 400\text{KSPS}$		2.6		mA
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 200\text{KSPS}$		1.3		mA
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 100\text{KSPS}$		0.65		mA
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 10\text{KSPS}$		64		μA
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 1\text{KSPS}$		6.4		μA
	Power-down supply current	$\overline{\text{CS}}/\text{SHDN} = V_{\text{DD}}$, SCLK Off		0.1		μA
		$\overline{\text{CS}}/\text{SHDN} = V_{\text{DD}}$, SCLK On		55		μA
P_{VDD}	Operating Power dissipation	$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 400\text{KSPS}$		13		mW
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 200\text{KSPS}$		6.5		mW
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 100\text{KSPS}$		3.25		mW
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 10\text{KSPS}$		320		μW
		$f_{\text{DCLOCK}} = 10\text{MHz}$, $F_{\text{S}} = 1\text{KSPS}$		32		μW
	Power dissipation in power-down	$\overline{\text{CS}}/\text{SHDN} = V_{\text{DD}}$, SCLK Off		0.5		μW
		$\overline{\text{CS}}/\text{SHDN} = V_{\text{DD}}$, SCLK On		275		μW
DIGITAL INPUT CHARACTERISTICS						
	Logic family		CMOS			
V_{IH}	Input high voltage		$0.7V_{\text{DD}}$		$V_{\text{DD}} + 0.3$	V
V_{IL}	Input low voltage		-0.3		$0.3V_{\text{DD}}$	V
C_{IN}	Input capacitance			5		pF
I_{IN}	Input current		-100		100	nA

Electrical Characteristics: VDD = +5V (continued)

At -40°C to $+125^{\circ}\text{C}$, $V_{\text{REF}} = +5\text{V}$, $-\text{IN} = \text{GND}$, $F_{\text{S}} = 400\text{kHz}$, and $f_{\text{DCLOCK}} = 25 \times F_{\text{S}}$, typical values are at $T_{\text{A}} = 25^{\circ}\text{C}$, unless otherwise noted.

DIGITAL OUTPUT CHARACTERISTICS						
	Logic family		CMOS			
	Data format		Straight binary			
V_{OH}		$I_{OH} = -100\mu A$	$V_{DD}-0.3$			V
V_{OL}		$I_{OL} = 100\mu A$			0.3	V
I_{OZ}	High-impedance state output current	$\overline{CS}/SHDN = V_{DD}, V_i = V_{DD} \text{ or } GND$	-100		100	nA
C_o	Output capacitance			5		pF
C_L	Load capacitance				30	pF

7.5 Electrical Characteristics: VDD = +2.7V

At $-40^\circ C$ to $+125^\circ C$, $V_{REF} = +2.5V$, $-IN = GND$, $F_S = 300kHz$, and $f_{DCLOCK} = 25 \times F_S$, typical values are at $T_A = 25^\circ C$, unless otherwise noted.⁽⁴⁾⁽⁵⁾

PARAMETER		CONDITIONS	MIN	TYP	MAX	UNIT
STATIC CONVERTER CHARACTERISTICS						
NMC	Resolution with no missing codes		16			Bits
INL	Integral non-linearity			± 1.5		LSB
DNL	Differential non-linearity			± 0.5		LSB
V_{OS}	Offset error			± 5		LSB
TCV_{OS}	Offset error drift			± 0.3		ppm/ $^\circ C$
GE	Gain error			± 8		LSB
TCGE	Gain error drift			± 0.4		ppm/ $^\circ C$
TUE	Total Unadjusted Error			± 5		LSB
	Noise			35		μV_{rms}
				7		LSB _{PP}
DYNAMIC CONVERTER CHARACTERISTICS⁽⁶⁾						
ENOB	Effective number of bits	$f_{IN}=2kHz$		14.33		Bits
		$f_{IN}=10kHz$		14.24		Bits
SINAD	Signal-to-noise plus distortion ratio	$f_{IN}=2kHz$		88		dB
		$f_{IN}=10kHz$		87.5		dB
SNR	Signal-to-noise ratio	$f_{IN}=2kHz$		88.5		dB
		$f_{IN}=10kHz$		88		dB
THD	Total harmonic distortion	$f_{IN}=2kHz$		-98		dB
		$f_{IN}=10kHz$		-97		dB
SFDR	Spurious-free dynamic range	$f_{IN}=2kHz$		100		dB
		$f_{IN}=10kHz$		98		dB
FPBW	60dB SINAD bandwidth	2.7V supply		700		kHz
PSRR	Power-supply rejection	$2.7V \leq V_{DD} \leq 3.6V$		0.5		LSB/V

(4) Data sheet minimum and maximum specification limits are ensured by design, test, or statistical analysis.

(5) Applies for 2.7V nominal supply: $V_{DD}(\min) = 2.7V$ and $V_{DD}(\max) = 3.6V$.

(6) All ac parameters are tested at -0.2 dBFS.

Electrical Characteristics: VDD = +2.7V (continued)

At $-40^\circ C$ to $+125^\circ C$, $V_{REF} = +2.5V$, $-IN = GND$, $F_S = 300kHz$, and $f_{DCLOCK} = 25 \times F_S$, typical values are at $T_A = 25^\circ C$, unless otherwise noted.

ANALOG INPUT CHARACTERISTICS

FSR	Full-scale range	+IN – (-IN)	0		V _{REF}	V
	Common-mode signal	-IN	-0.3		0.5	V
C _{IN}	Input capacitance	-IN = GND, during sampling		48		pF
I _{IL}	Input leakage current	$\overline{CS}/SHDN=V_{DD}$, SCLK off		250		nA
REFERENCE INPUT CHARACTERISTICS						
V _{REF}	Reference voltage		0.1		V _{DD}	V
C _{REF}	Reference input capacitance			48		pF
I _{REF}	Reference input current	F _S =300KSPS		67		μA
		F _S =250KSPS		56		μA
		F _S =100KSPS		23		μA
		F _S =10KSPS		3		μA
		$\overline{CS}/SHDN=V_{DD}$		0.1		μA
SAMPLING DYNAMIC CHARACTERISTICS						
t _{CONV}	Conversion time(18 DCLOCKS)	25kHz ≤ f _{DCLOCK} ≤ 7.5MHz	18			T _{DCLOCK}
t _{AQ}	Acquisition time(5 DCLOCKS)		4.5	5		T _{DCLOCK}
F _S	Throughput rate(25 DCLOCKS)				300	KSPS
f _{DCLOCK}	Clock frequency		0.025		7.5	MHz
POWER SUPPLY CHARACTERISTICS						
I _{VDD}	Operating supply current	f _{DCLOCK} =7.5MHz, F _S =300KSPS		1.8		mA
		f _{DCLOCK} =7.5MHz, F _S =200KSPS		1.2		mA
		f _{DCLOCK} =7.5MHz, F _S =100KSPS		0.6		mA
		f _{DCLOCK} =7.5MHz, F _S =10KSPS		60		μA
		f _{DCLOCK} =7.5MHz, F _S =1KSPS		6		mA
	Power-down supply current	$\overline{CS}/SHDN=V_{DD}$, SCLK Off		0.1		μA
		$\overline{CS}/SHDN=V_{DD}$, SCLK On		5		μA
P _{VDD}	Operating Power dissipation	f _{DCLOCK} =7.5MHz, F _S =300KSPS		4.86		mW
		f _{DCLOCK} =7.5MHz, F _S =200KSPS		3.24		mW
		f _{DCLOCK} =7.5MHz, F _S =100KSPS		1.62		mW
		f _{DCLOCK} =7.5MHz, F _S =10KSPS		162		μW
		f _{DCLOCK} =7.5MHz, F _S =1KSPS		16.2		μW
	Power dissipation in power-down	$\overline{CS}/SHDN=V_{DD}$, SCLK Off		0.27		μW
		$\overline{CS}/SHDN=V_{DD}$, SCLK On		13.5		μW

Electrical Characteristics: VDD = +2.7V (continued)

At -40°C to +125°C, V_{REF} = +2.5V, -IN = GND, F_S = 300kHz, and f_{DCLOCK} = 25 × F_S, typical values are at T_A = 25°C, unless otherwise noted.

DIGITAL INPUT CHARACTERISTICS		
	Logic family	CMOS

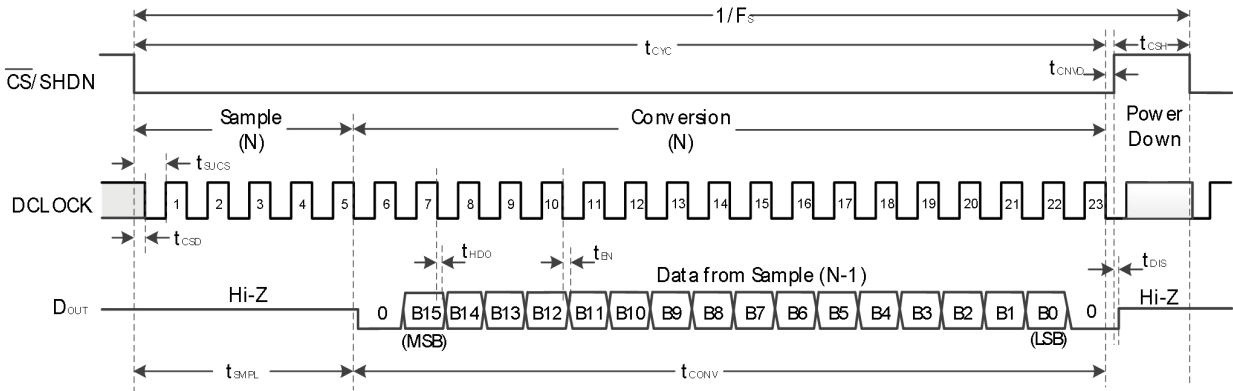
V _{IH}	Input high voltage		0.7V _{DD}		V _{DD} +0.3	V
V _{IL}	Input low voltage		-0.3		0.3V _{DD}	V
C _{IN}	Input capacitance			5		pF
I _{IN}	Input current		-50		50	nA
DIGITAL OUTPUT CHARACTERISTICS						
	Logic family		CMOS			
	Data format		Straight binary			
V _{OH}		I _{OH} = -100μA	V _{DD} -0.3			V
V _{OL}		I _{OL} = 100μA			0.3	V
I _{OZ}	High-impedance state output current	$\overline{CS}/SHDN = V_{DD}, V_I = V_{DD} \text{ or } GND$	-50		50	nA
C _O	Output capacitance			5		pF
C _L	Load capacitance				30	pF

7.6 Timing Requirements

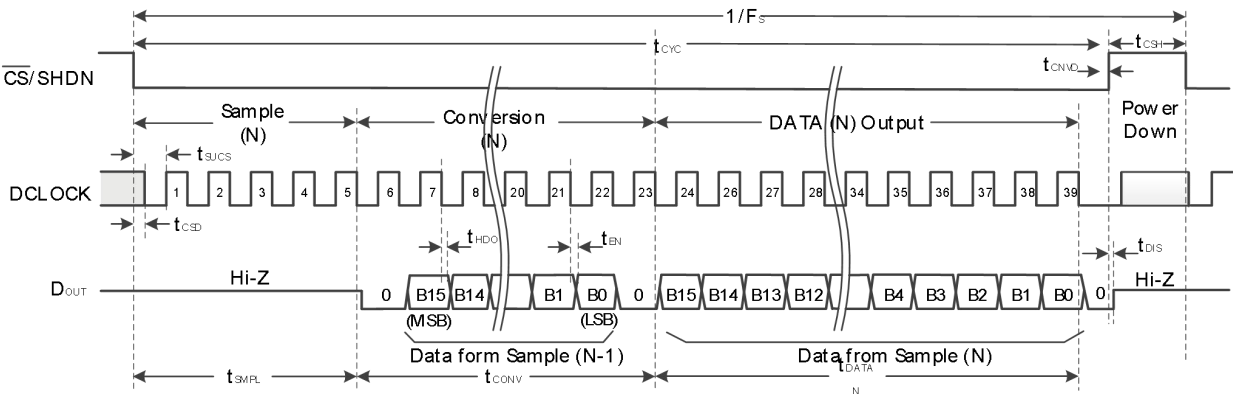
-40°C ≤ T_A ≤ 125°C, V_{DD} = 2.7 V to 5.5 V (unless otherwise noted) ⁽¹⁾

PARAMETER		CONDITIONS	MIN	TYP	MAX	UNIT
f _{DCLOCK}	DCLOCK frequency	V _{DD} = 4.5V to 5.5V			10	MHz
		V _{DD} = 2.7V to 3.6V			7.5	MHz
F _S	Throughput rate	V _{DD} = 4.5V to 5.5V			400	KSPS
		V _{DD} = 2.7V to 3.6V			300	KSPS
t _{SMPL}	Analog input sample time		4.5		5	DCLOCKS
t _{CONV}	Conversion time		18			DCLOCKS
t _{CYC}	Complete cycle time		25			DCLOCKS
t _{CSH}	Minimum CS pulse width		10			ns
t _{CSD}	CS falling to DCLOCK low				0	ns
t _{SUCS}	CS falling to DCLOCK rising		20			ns
t _{HDO}	DCLOCK falling to current D _{OUT} not valid			15		ns
t _{DIS}	CS rising to D _{OUT} tri-state				100	ns
t _{EN}	DCLOCK falling to D _{OUT} enabled				50	ns
t _{CNVD}	The 23rd DCLOCK falling to CS rising		50			ns
t _F	D _{OUT} fall time				40	ns
t _R	D _{OUT} rise time				40	ns
t _{WH}	Pulse duration, DCLOCK high		40			ns
t _{WL}	Pulse duration, DCLOCK low		40			ns
t _{AD}	Aperture delay			7.5		ns
t _{AJ}	Aperture jitter			30		ns

(1) Measured with 50pF load.



NOTE: (1) A minimum of 24 clock cycles are required for 16-bit conversion; 25 clock cycles are shown. If $\overline{CS}/SHDN$ remains low at the end of conversion, a new data stream from sample (N) is shifted out.



NOTE: (2) After completing the data transfer, if further clocks are applied with CS low, the A/D converter will output data(n) repeatedly.

Figure 1. ZMB1430B Serial Interface Timing Diagram

7.7 Typical Characteristics: VDD = +5V

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

T_A = 25°C, V_{DD} = +5 V, V_{REF} = +5 V, F_S = 400 KSPS, f_{DLOCK} = 10 MHz, f_{IN} = 2.07 kHz, P_{IN} = -0.2 dBFS (unless otherwise noted).

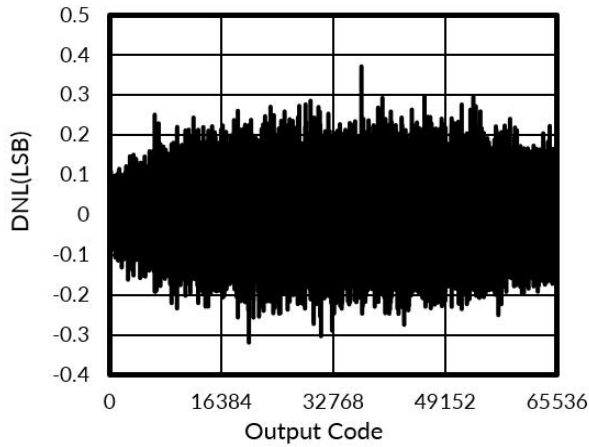


Figure 2. DNL vs Output Code

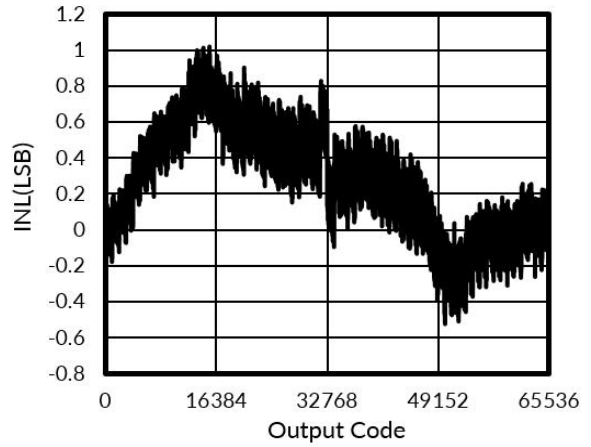


Figure 3. INL vs Output Code

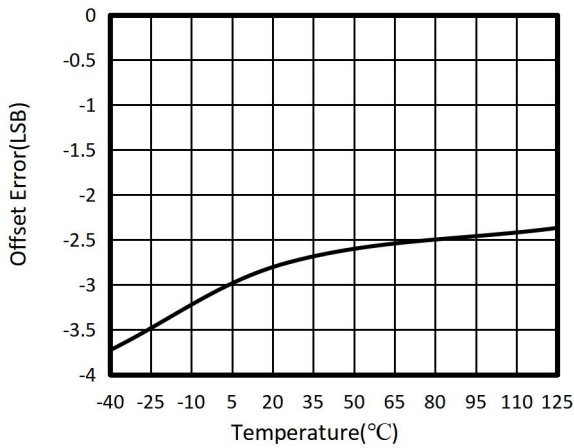


Figure 4. Offset Error vs Temperature

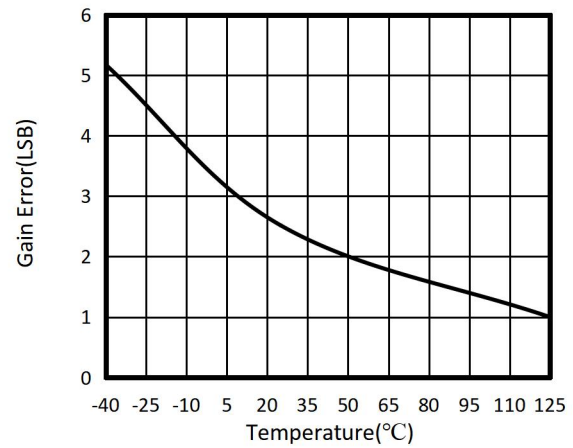


Figure 5. Gain Error vs Temperature

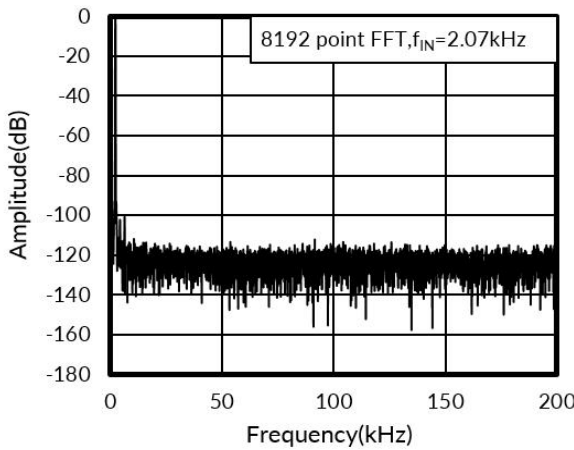


Figure 6. Spectral Response

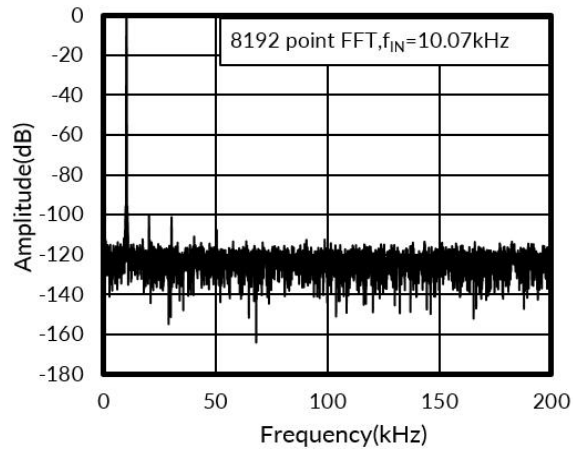


Figure 7. Spectral Response

Typical Characteristics: VDD = +5V (continued)

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

T_A = 25°C, V_{DD} = +5 V, V_{REF} = +5 V, F_S = 400 KSPS, f_{DCLK} = 10 MHz, f_{IN} = 2.07 kHz, P_{IN} = -0.2 dBFS (unless otherwise noted).

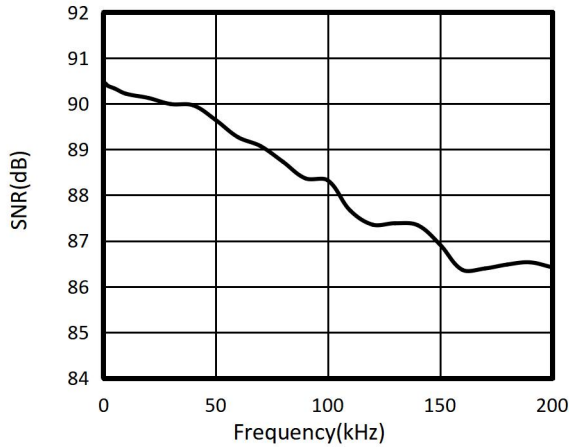


Figure 8. SNR vs Input Frequency

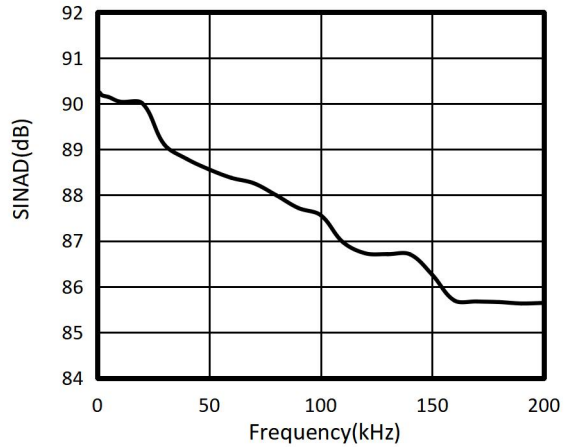


Figure 9. SINAD vs Input Frequency

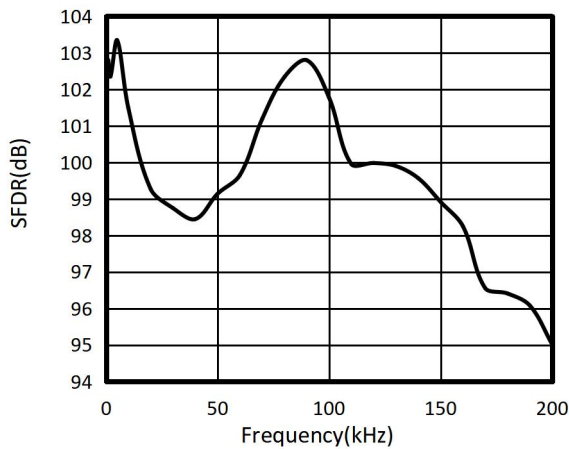


Figure 10. SFDR vs Input Frequency

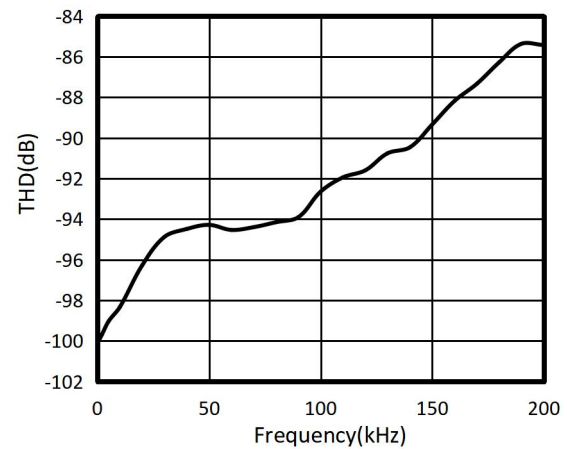


Figure 11. THD vs Input Frequency

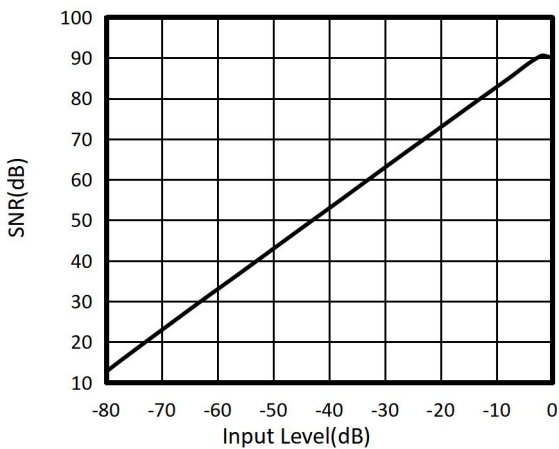


Figure 12. SNR vs Input Level

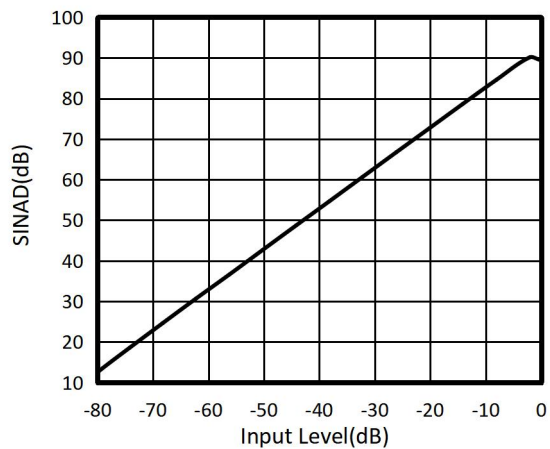


Figure 13. SINAD vs Input Level

Typical Characteristics: VDD = +5V (continued)

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

T_A = 25°C, V_{DD} = +5 V, V_{REF} = +5 V, F_S = 400 KSPS, f_{DLOCK} = 10 MHz, f_{IN} = 2.07 kHz, P_{IN} = -0.2 dBfs (unless otherwise noted).

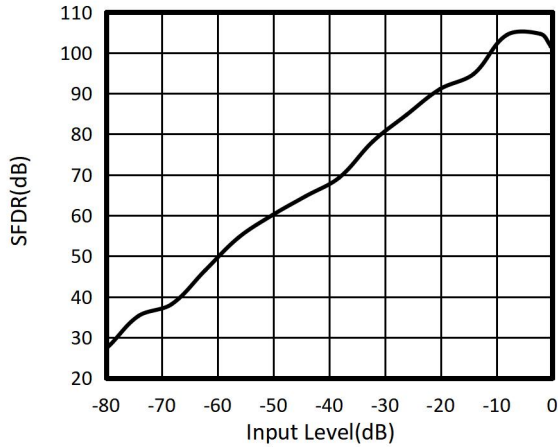


Figure 14. SFDR vs Input Level

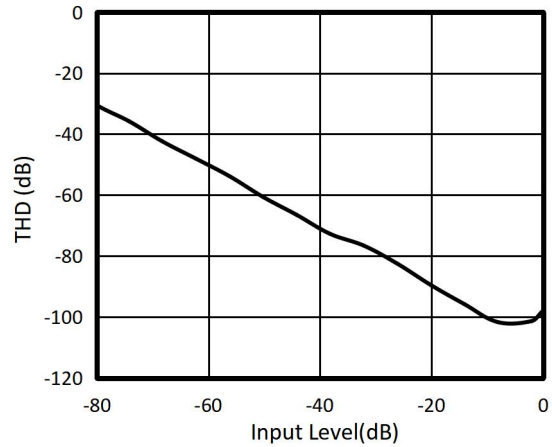


Figure 15. THD vs Input Level

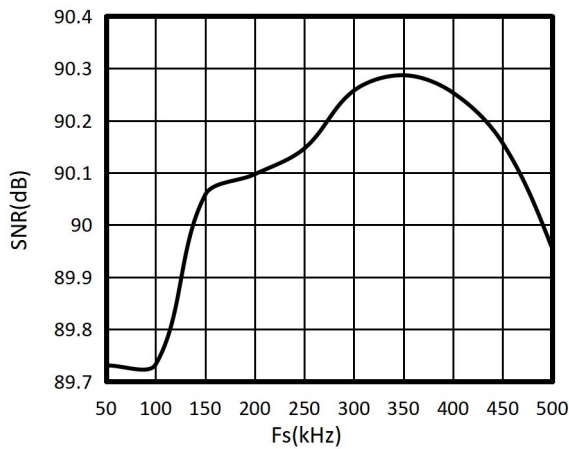


Figure 16. SNR vs Fs

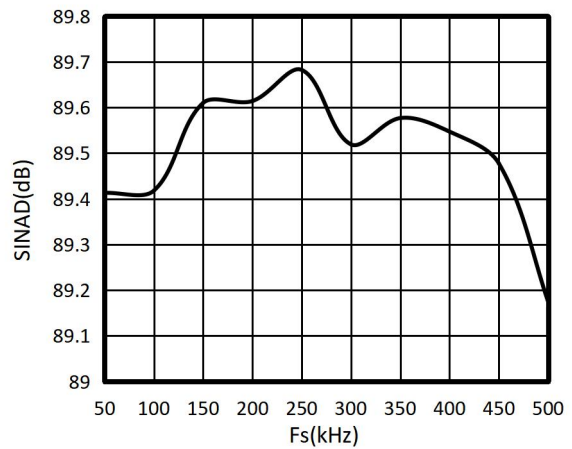


Figure 17. SINAD vs Fs

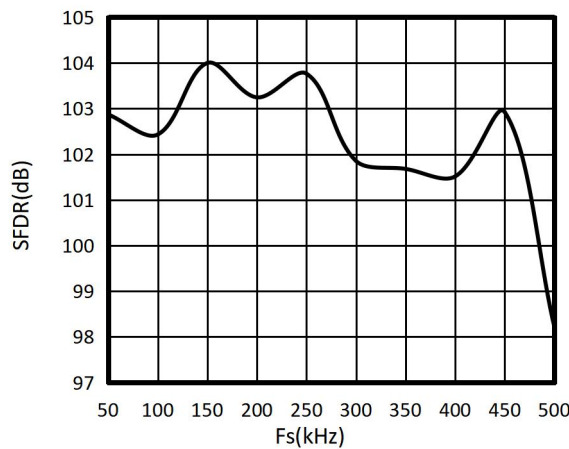


Figure 18. SFDR vs Fs

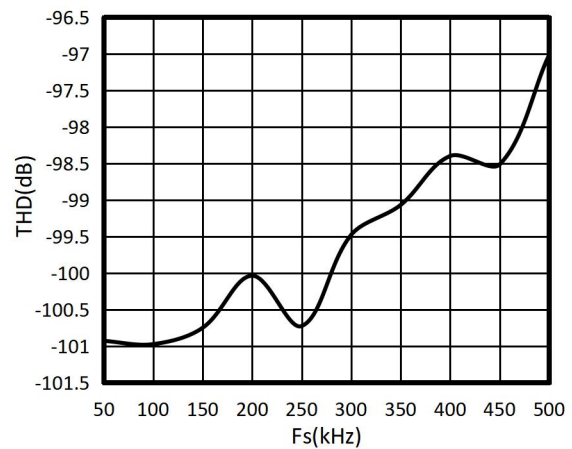


Figure 19. THD vs Fs

Typical Characteristics: VDD = +5V (continued)

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

T_A = 25°C, V_{DD} = +5 V, V_{REF} = +5 V, F_S = 400 KSPS, f_{DCLK} = 10 MHz, f_{IN} = 2.07 kHz, P_{IN} = -0.2 dBfs (unless otherwise noted).

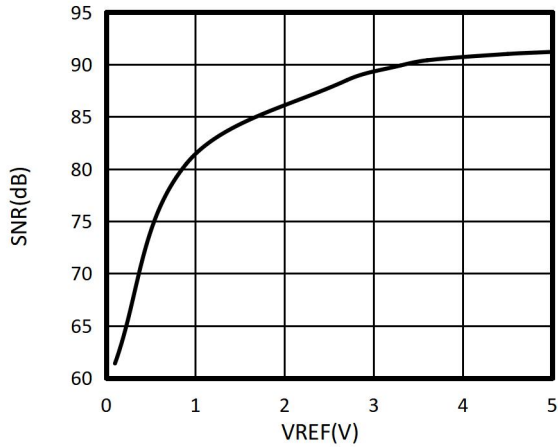


Figure 20. SNR vs V_{REF}

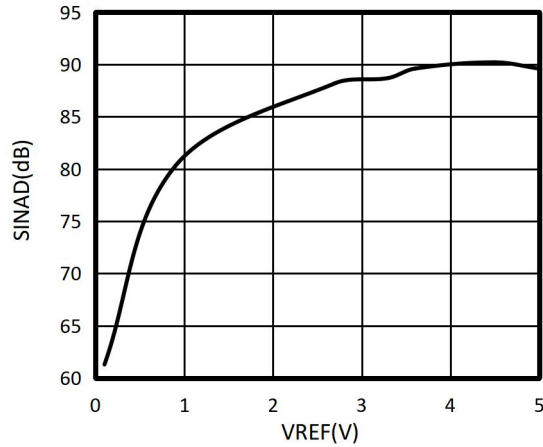


Figure 21. SINAD vs V_{REF}

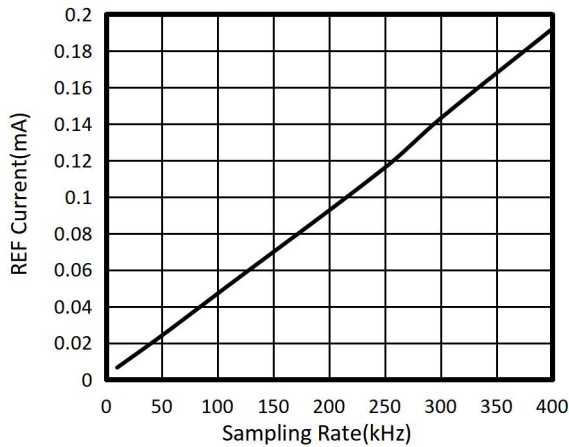


Figure 22. REF Current vs Sampling Rate

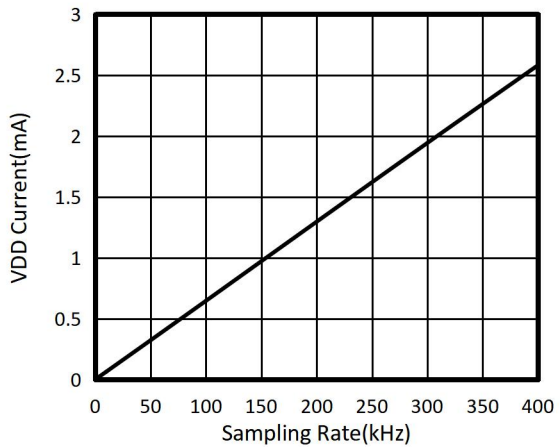


Figure 23. VDD Current vs Sampling Rate

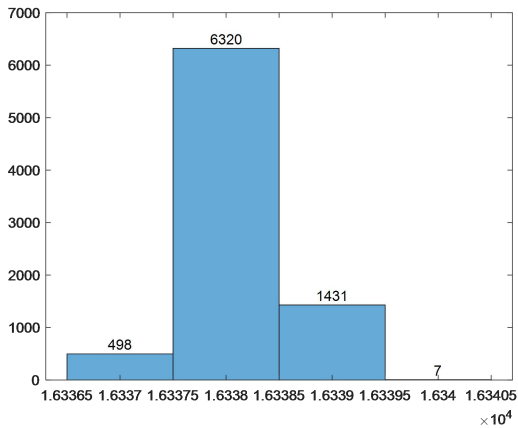


Figure 24. Output Code Histogram from a DC Input (8192 Conversions)

7.8 Typical Characteristics: VDD = +2.7V

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

T_A = 25°C, V_{DD} = +2.7 V, V_{REF} = +2.5 V, F_S = 300 KSPS, f_{DCLOCK} = 7.5 MHz, f_{IN} = 2.07 kHz, P_{IN} = -0.2 dBFS (unless otherwise noted).

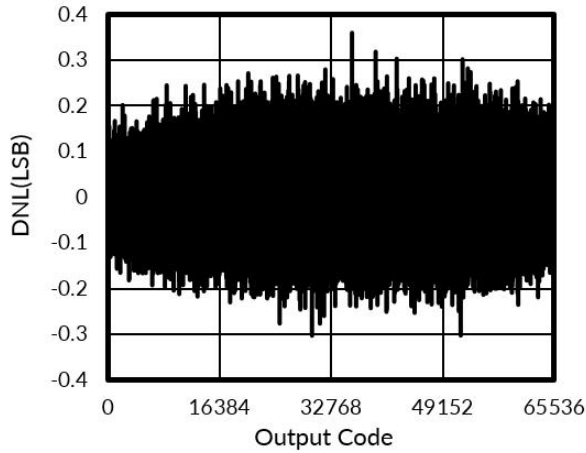


Figure 25. DNL vs Output Code

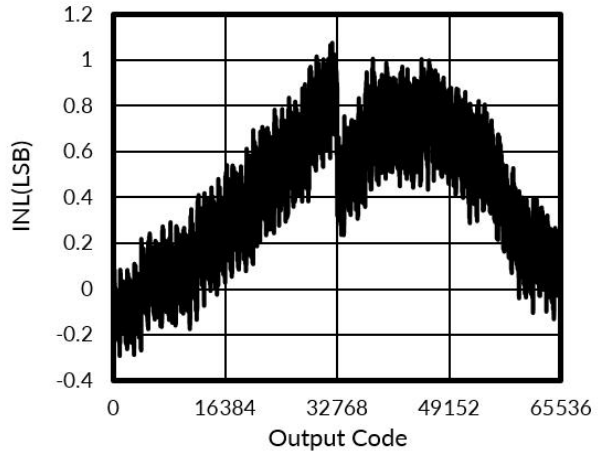


Figure 26. INL vs Output Code

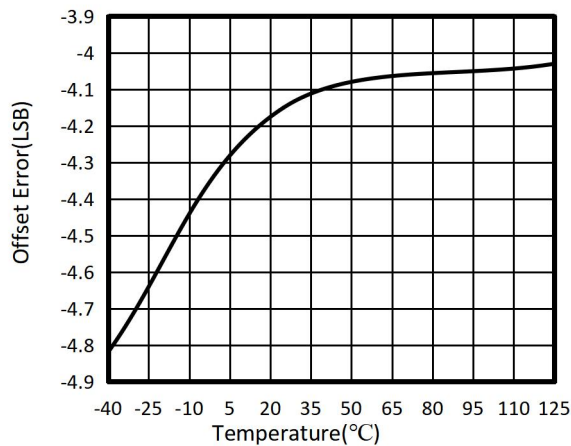


Figure 27. Offset Error vs Temperature

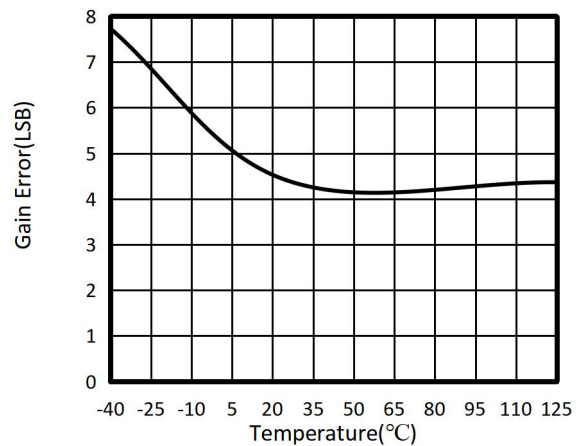


Figure 28. Gain error vs Temperature

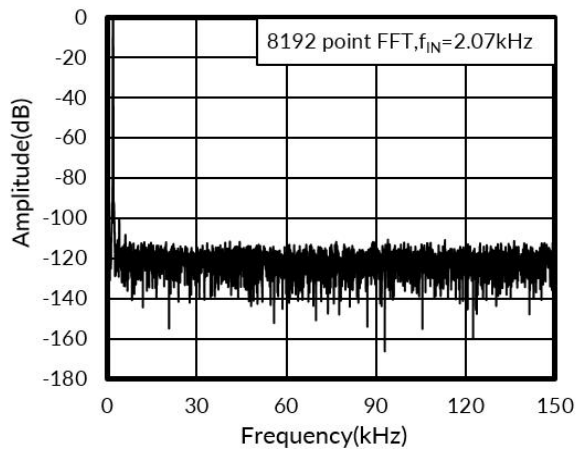


Figure 29. Spectral Response

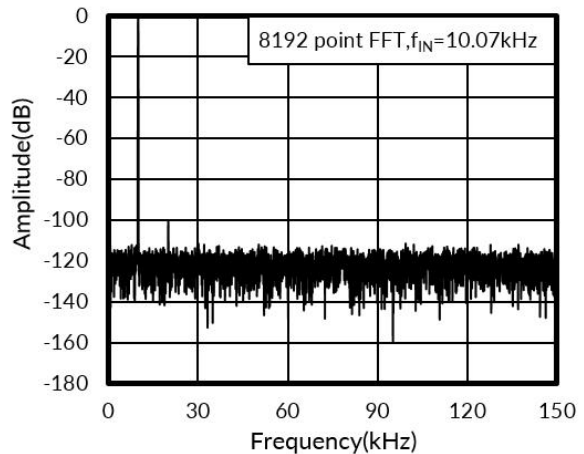


Figure 30. Spectral Response

Typical Characteristics: VDD = +2.7V (continued)

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

$T_A = 25^\circ\text{C}$, $V_{DD} = +2.7\text{ V}$, $V_{REF} = +2.5\text{ V}$, $F_S = 300\text{ KSPS}$, $f_{DCLOCK} = 7.5\text{ MHz}$, $f_{IN} = 2.07\text{ kHz}$, $P_{IN} = -0.2\text{ dBFs}$ (unless otherwise noted).

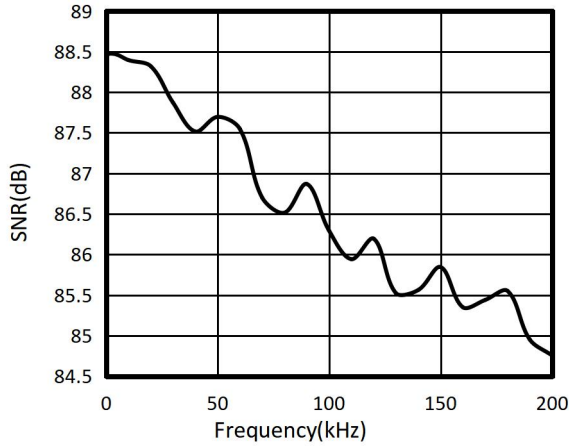


Figure 31. SNR vs Input Frequency

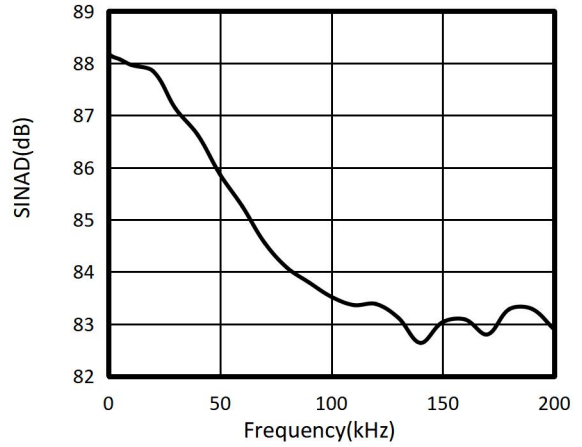


Figure 32. SINAD vs Input Frequency

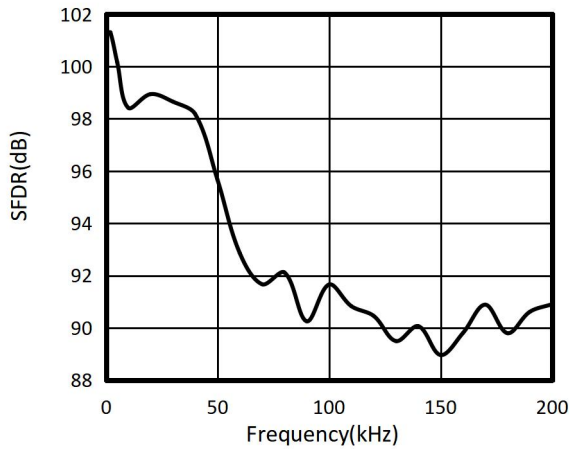


Figure 33. SFDR vs Input Frequency

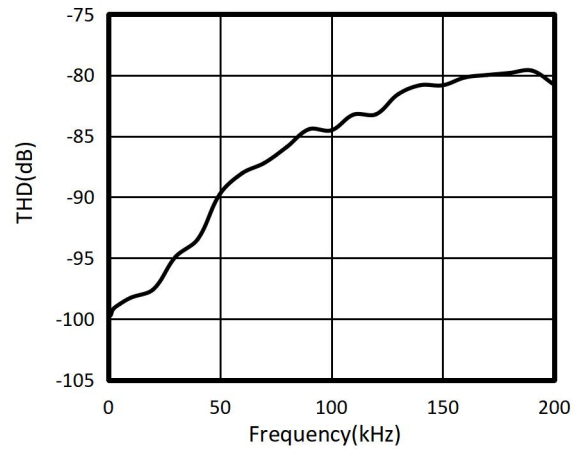


Figure 34. THD vs Input Frequency

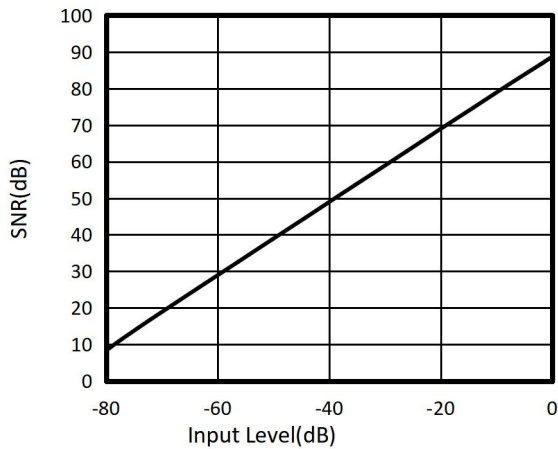


Figure 35. SNR vs Input Level

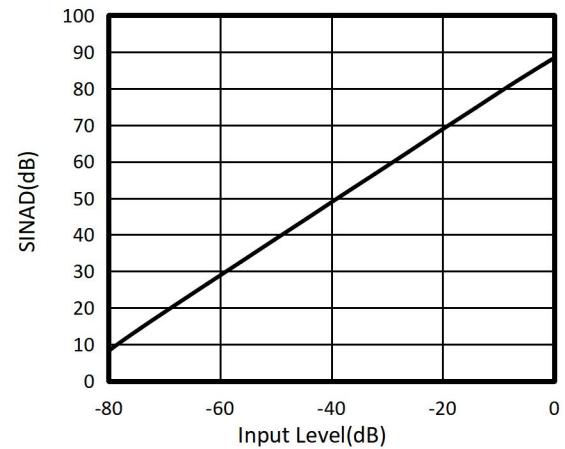


Figure 36. SINAD vs Input Level

Typical Characteristics: VDD = +2.7V (continued)

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

T_A = 25°C, V_{DD} = +2.7 V, V_{REF} = +2.5 V, F_S = 300 KSPS, f_{DCLOCK} = 7.5 MHz, f_{IN} = 2.07 kHz, P_{IN} = -0.2 dBFS (unless otherwise noted).

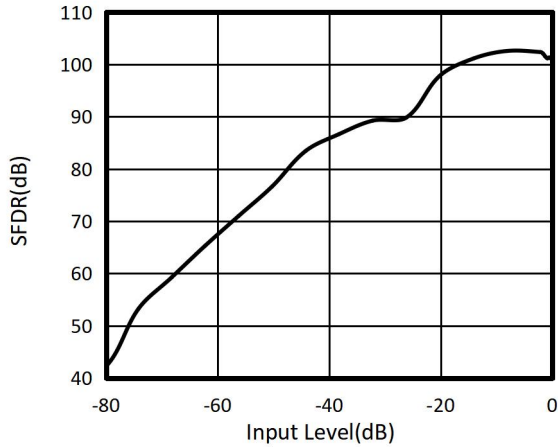


Figure 37. SFDR vs Input Level

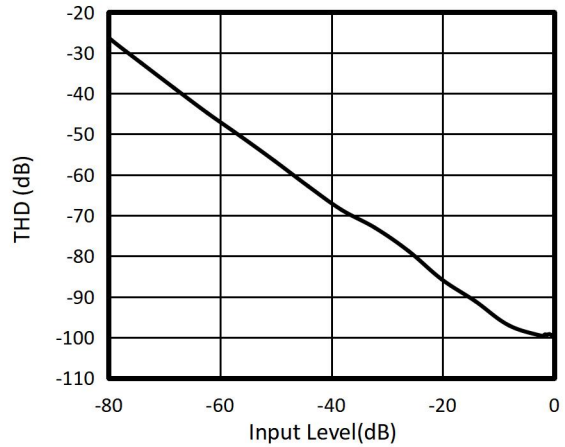


Figure 38. THD vs Input Level

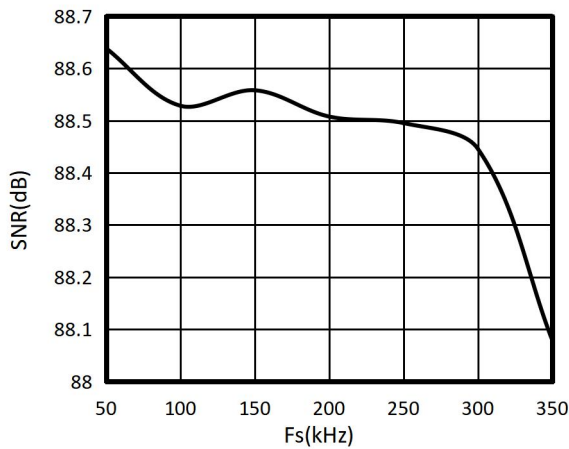


Figure 39. SNR vs Fs

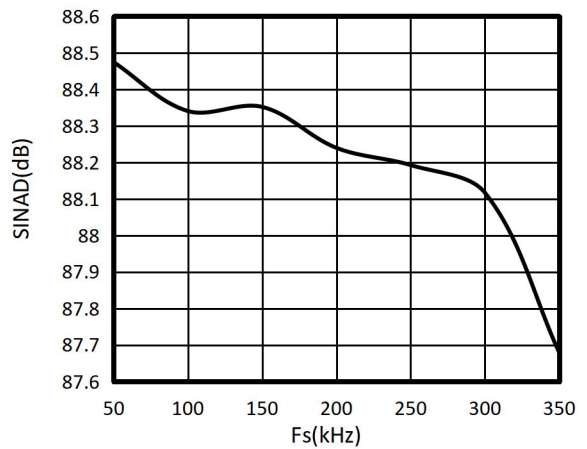


Figure 40. SINAD vs Fs

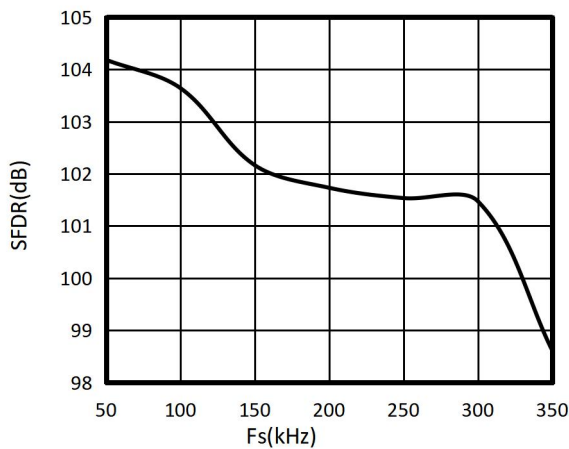


Figure 41. SFDR vs Fs

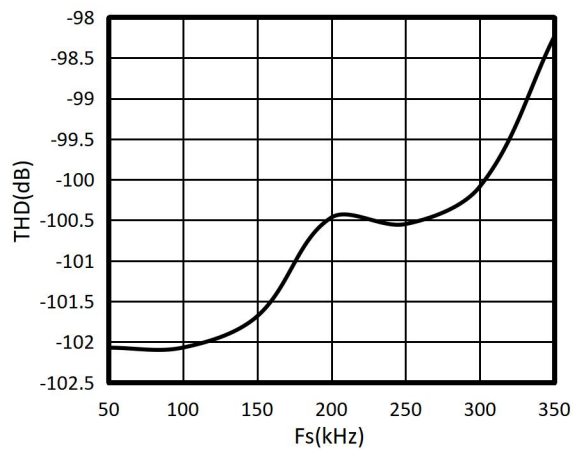


Figure 42. THD vs Fs

Typical Characteristics: VDD = +2.7V (continued)

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

T_A = 25°C, V_{DD} = +2.7 V, V_{REF} = +2.5 V, F_S = 300 KSPS, f_{DCLOCK} = 7.5 MHz, f_{IN} = 2.07 kHz, P_{IN} = -0.2 dBFS (unless otherwise noted).

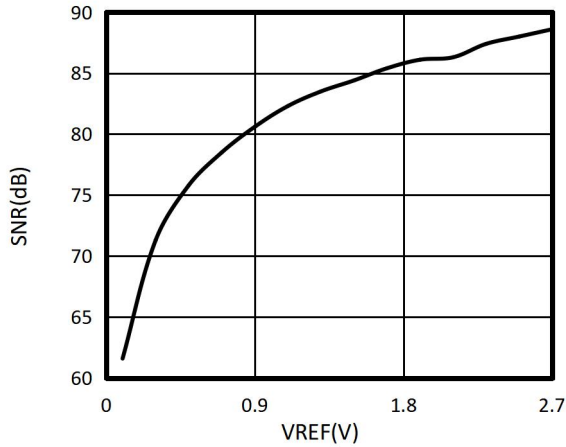


Figure 43. SNR vs V_{REF}

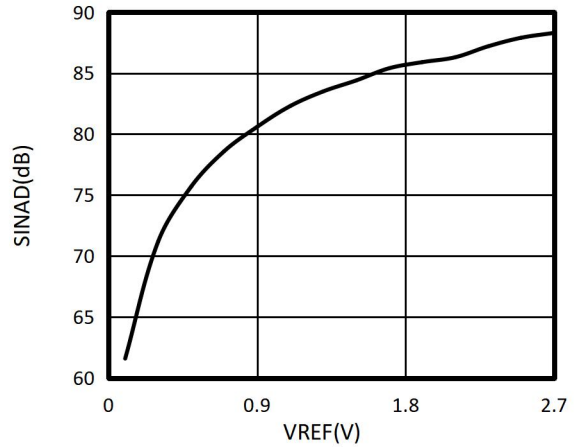


Figure 44. SINAD vs V_{REF}

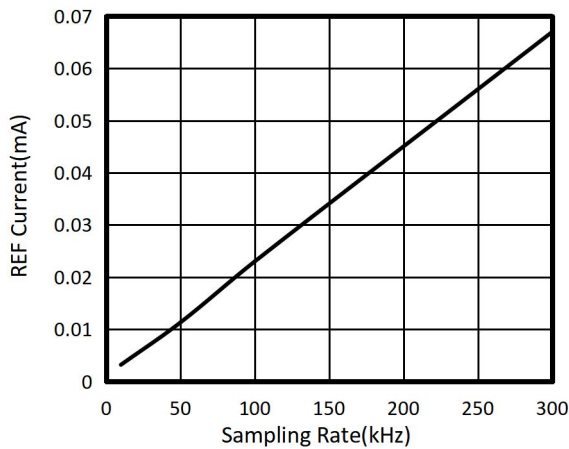


Figure 45. REF Current vs Sampling Rate

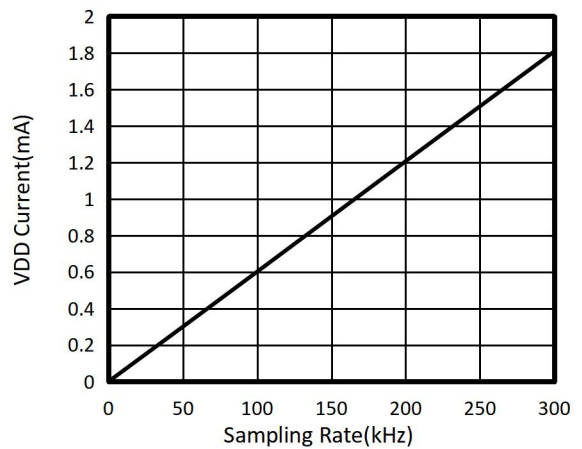


Figure 46. VDD Current vs Sampling Rate

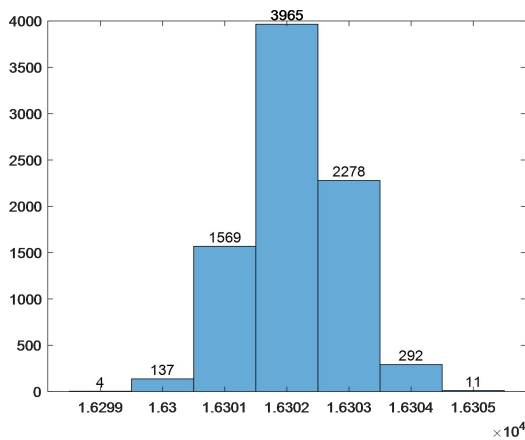


Figure 47. Output Code Histogram from a DC Input (8192 Conversions)

8 Detailed Description

8.1 Overview

The ZMB1430B device is a classic successive approximation register (SAR) analog-to-digital (A/D) converter. The architecture is based on capacitive redistribution, which inherently includes a sample and hold function. The architecture and process allow the ZMB1430B to acquire and convert an analog signal at up to 400,000 conversions per second while consuming less than 13mW from V_{DD} .

Differential linearity for the ZMB1430B is factory-adjusted via a package-level trim procedure. The state of the trim elements is stored in non-volatile memory and is continuously updated after each acquisition cycle, just prior to the start of the successive approximation operation. This process ensures that one complete conversion cycle always returns the part to its factory-adjusted state in the event of a power interruption.

The ZMB1430B requires an external reference, an external clock, and a single power source (V_{DD}). The external reference can be any voltage between 0.1 V and V_{DD} . The value of the reference voltage directly sets the range of the analog input. The reference input current depends on the conversion rate of the ZMB1430B.

The external clock can vary between 25 kHz (1-kHz throughput) and 10 MHz (400-kHz throughput). The duty cycle of the clock is essentially unimportant, as long as the minimum high and low times are at least 40 ns ($V_{DD} = 2.7$ V or greater). The minimum clock frequency is set by the leakage on the internal capacitors to the ZMB1430B.

The analog input is provided to two input pins: +IN and -IN. When a conversion is initiated, the pseudo-differential input on these pins is sampled on the internal capacitor array. While a conversion is in progress, both inputs are disconnected from any internal function.

The digital result of the conversion is clocked out by the DCLOCK input and is provided serially, most significant bit first, on the D_{OUT} pin. The digital data that is provided on the D_{OUT} pin is for the previous conversion. It is possible to continue to clock the ZMB1430B after the conversion is complete and to obtain the serial data from the conversion currently in progress. See the Timing Information section for more information.

8.2 Analog Input

The analog input of ZMB1430B is pseudo-differential. The +IN and -IN input pins allow for a pseudo-differential input signal. The amplitude of the input is the difference between the +IN and -IN input, or $(+IN) - (-IN)$. Unlike some converters of this type, the -IN input is not resampled later in the conversion cycle. When the converter goes into Hold mode or conversion, the voltage difference between +IN and -IN is captured on the internal capacitor array.

The range of the -IN input is limited to -0.3V to +0.5V. As a result of this limitation, the pseudo-differential input could be used to reject signals that are common to both inputs in the specified range. Thus, the -IN input is best used to sense a remote signal ground that may move slightly with respect to the local ground potential.

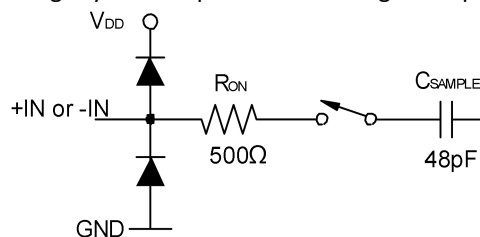


Figure 48. Equivalent Analog Input Circuit of ZMB1430B

The general method for driving the analog input of the ZMB1430B is shown in **Figure 48** and **Figure 49**. The -IN input is held at the common-mode voltage. The +IN input swings from -IN (or common-mode voltage) to -IN + V_{REF} (or common-mode voltage + V_{REF}), and the peak-to-peak amplitude is $+V_{REF}$. The value of V_{REF} determines the range over which the common-mode voltage may vary, as shown in **Figure 50**.

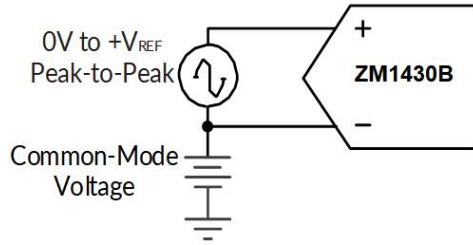


Figure 49. Methods of Driving the ZMB1430B

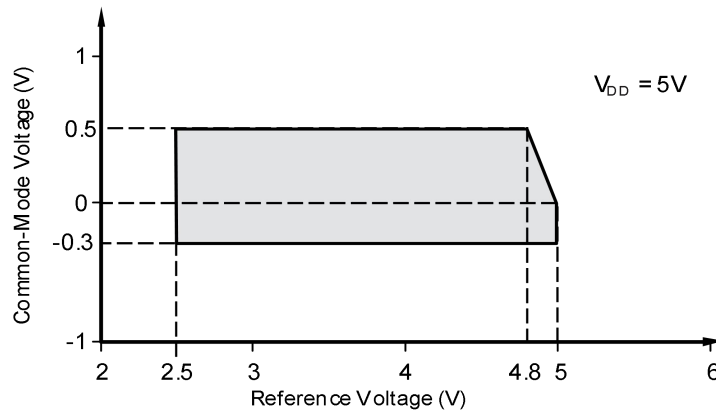
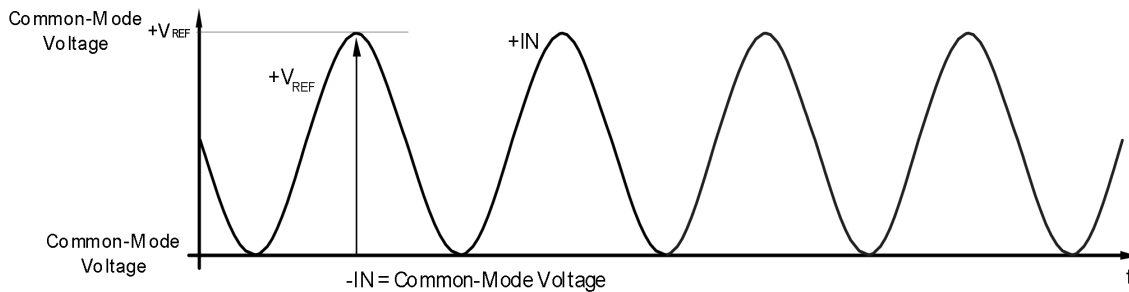


Figure 50. -IN Analog Input: Common-Mode Voltage Range vs V_{REF}



NOTE: The maximum differential voltage between +IN and -IN of the ZMB1430B is V_{REF} . See Figure 50 for a further explanation of the common-mode voltage range for pseudo-differential inputs.

Figure 51. Pseudo-differential Input Mode of the ZMB1430B

The input current required by the analog inputs depends on a number of factors: sample rate, input voltage, source impedance, and power-down mode. Essentially, the current into the ZMB1430B charges the internal capacitor array during the sample period. After this capacitance has been fully charged, there is no further input current. The source of the analog input voltage must be able to charge the input capacitance (48pF) to a 16-bit settling level within 4.5 clock cycles (0.45 μ s). When the converter goes into Hold mode, or while it is in Power-Down mode, the input impedance is greater than 10M Ω .

Care must be taken regarding the absolute analog input voltage. To maintain the linearity of the converter, the -IN input should not drop below GND-0.3V or exceed GND+0.5V. The +IN input should always remain within the range of GND-0.3V to VDD+0.3V, or -IN to -IN+VREF, whichever limit is reached first. Outside of these ranges, the converter linearity may not meet specifications. To minimize noise, low bandwidth input signals with low-pass filters should be used. In each case, care should be taken to ensure that the output impedance of the sources driving the +IN and -IN inputs are matched. Often, a small capacitor between the positive and negative inputs helps to match their impedance. To obtain maximum performance from the ZMB1430B, the input circuit from **Figure 52** is recommended.

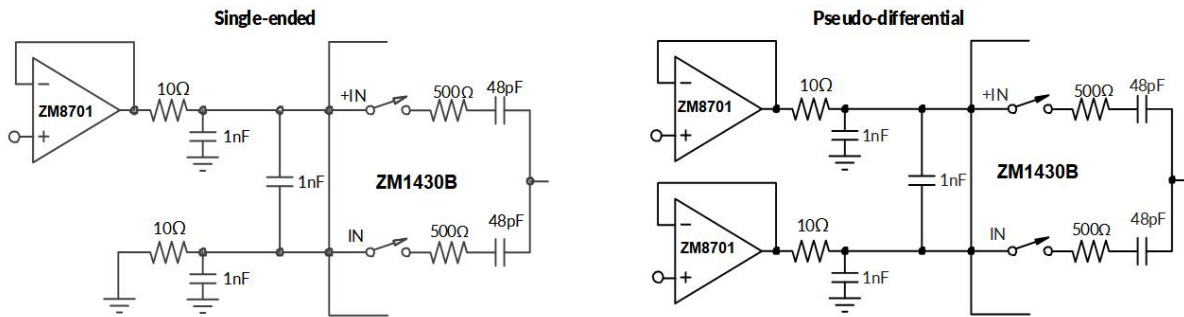


Figure 52. Single-ended and Pseudo-differential Methods of Interfacing the ZMB1430B

8.3 Reference Input

The external reference sets the analog input range. The ZMB1430B operates with a reference in the range of 0.1V to VDD. There are several important implications to this.

As the reference voltage is reduced, the analog voltage weight of each digital output code is reduced. This is often referred to as the least significant bit (LSB) size and is equal to the reference voltage divided by 65,536. This means that any offset or gain error inherent in the A/D converter will appear to increase (in terms of LSB size) as the reference voltage is reduced. For a reference voltage of 2.5V, the value of the LSB is 38.15μV, and for a reference voltage of 5V, the LSB is 76.3μV.

The noise inherent in the converter will also appear to increase with a lower LSB size. With a 5V reference, the internal noise of the converter typically contributes only 4LSB peak-to-peak of potential error to the output code. When the external reference is 2.5V, the potential error contribution from the internal noise will be two times larger (7LSB). The errors arising from the internal noise are Gaussian in nature and can be reduced by averaging consecutive conversion results.

With lower reference voltages, extra care should be taken to provide a clean layout including adequate bypassing, a clean power supply, a low-noise reference, and a low-noise input signal. Due to the lower LSB size, the converter is also more sensitive to external sources of error, such as nearby digital signals and electromagnetic interference.

The equivalent input circuit for the reference voltage is presented in **Figure 53**. During the conversion process, an equivalent capacitor of 48pF is switched on. To obtain optimum performance from the ZMB1430B, special care must be taken in designing the interface circuit to the reference input pin. To ensure a stable reference voltage, a 47μF tantalum capacitor with low ESR should be connected as close as possible to the input pin. If a high output impedance reference source is used, an additional operational amplifier with a current-limiting resistor must be placed in front of the capacitors.

When the ZMB1430B is in Power-Down mode, the input resistance of the reference pin will have a value of 10MΩ. Since the input capacitors must be recharged before the next conversion starts, an operational amplifier with good dynamic characteristics must be used to buffer the reference input.

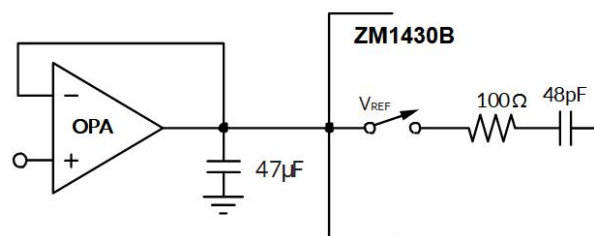


Figure 53. Input Reference Circuit and Interface

8.4 Noise

The transition noise of the ZMB1430B itself is extremely low, as shown in **Figure 24** (+5V) and **Figure 47** (+2.7V); it is much lower than competing A/D converters. These histograms were generated by applying a low-noise DC input and initiating 8192 conversions. The digital output of the A/D converter will vary in output code because of the internal noise of the ZMB1430B. Using a histogram to plot the output codes, the distribution should appear bell-shaped with the peak of the bell curve representing the nominal code for the input value. The $\pm 1\sigma$, $\pm 2\sigma$, and $\pm 3\sigma$ distributions will represent 68.3%, 95.5%, and 99.7%, respectively, of all codes. The transition noise can be calculated by dividing the number of codes measured by 6, which yields the $\pm 3\sigma$ distribution, or 99.7%, of all codes. Statistically, up to three codes could fall outside the distribution when executing 1000 conversions. The ZMB1430B, with < 4 output codes for the $\pm 3\sigma$ distribution, yields $< \pm 0.6\text{LSB}$ of transition noise. Remember, to achieve this low-noise performance, the peak-to-peak noise of the input signal and reference must be $< 50\mu\text{V}$.

8.5 Signal Levels

The ZMB1430B has a wide range of power-supply voltage. The A/D converter, as well as the digital interface circuit, is designed to accept and operate from 2.7V up to 5.5V. This voltage range will accommodate different logic levels. When the ZMB1430B power-supply voltage is in the range of 4.5V to 5.5V (5V logic level), the ZMB1430B can be connected directly to another 5V, CMOS integrated circuit. When the ZMB1430B power-supply voltage is in the range of 2.7V to 3.6V (3V logic level), the ZMB1430B can be connected directly to another 3.3V LVCMOS integrated circuit.

9 Digital Interface

9.1 Serial Interface

The ZMB1430B communicates with microprocessors and other digital systems via a synchronous 3-wire serial interface, as illustrated in the Timing Information section. The DCLOCK signal synchronizes the data transfer, with each bit being transmitted on the falling edge of DCLOCK. Most receiving systems will capture the bitstream on the rising edge of DCLOCK. However, if the minimum hold time for DOUT is acceptable, the system can use the falling edge of DCLOCK to capture each bit.

A falling CS signal initiates the conversion and data transfer. The first 4.5 to 5.0 clock periods of the conversion cycle are used to sample the input signal. After the fifth falling DCLOCK edge, DOUT is enabled and will output a low value for one clock period. For the next 16 DCLOCK periods, DOUT will output the previous conversion result, most significant bit first. After the least significant bit (B0) has been output, subsequent clocks will output the current conversion result with a low value as starting point. After the least significant bit (B0) has been exported. Subsequent clocks will repeat the current conversion result with a low value as starting point.

A new conversion is initiated only when CS has been taken high and returned low.

9.2 Data Format

The output data from the ZMB1430B is in Straight Binary format, as shown in **Figure 54**. This figure represents the ideal output code for a given input voltage and does not include the effects of offset, gain error, or noise.

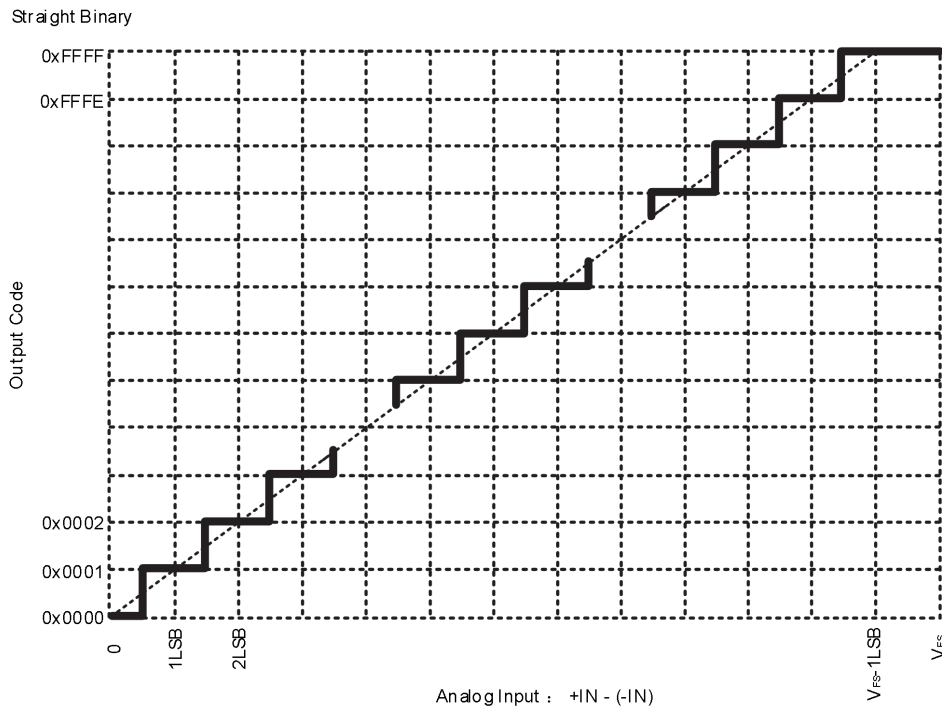


Figure 54. ZMB1430B Ideal Transfer Characteristic

10 Power Dissipation

The architecture of the converter, the semiconductor fabrication process, and a careful design allow the ZMB1430B to convert at up to a 400kHz rate while requiring very little power. However, for the absolute lowest power dissipation, there are several things to keep in mind.

The power dissipation of the ZMB1430B scales directly with conversion rate. Therefore, the first step to achieving the lowest power dissipation is to find the lowest conversion rate that will satisfy the requirements of the system.

In addition, the ZMB1430B goes into Power-Down mode under two conditions: when the conversion is complete and whenever CS is high (see the Timing Information section). Ideally, each conversion should occur as quickly as possible, preferably at a 10MHz clock rate. This way, the converter spends the longest possible time in Power-Down mode. This is very important because the converter not only uses power on each DCLOCK transition (as is typical for digital CMOS components), but also uses some current for the analog circuitry, such as the comparator. The analog section dissipates power continuously until Power-Down mode is entered.

Figure 22 and **Figure 23** (+5V), and **Figure 45** and **Figure 46** (+2.7V) illustrate the current consumption of the ZMB1430B versus sample rate. For these graphs, the converter is clocked at maximum speed regardless of the sample rate. CS is held high during the remaining sample period.

There is an important distinction between the power-down mode that is entered after a conversion is complete and the full power-down mode that is enabled when CS is high. CS low will only shut down the analog section. The digital section is completely shut down only when CS is high. Thus, if CS is left low at the end of a conversion, and the converter is continually clocked, the power consumption will not be as low as when CS is high.

11 Application and Implementation

Figure 55 and Figure 56 show two examples of a basic data acquisition system. The 5Ω resistor and 0.1μF to 10μF capacitor filters the microcontroller noise on the supply, as well as any high-frequency noise from the supply itself. The exact values should be picked such that the filter provides adequate rejection of noise.

The 100Ω resistors serial on CS and DCLOCK are used to filter out the digital overshoot, respectively. The exact values should be selected based on the conversion speed, rising/falling time of CS and DCLOCK, and so on.

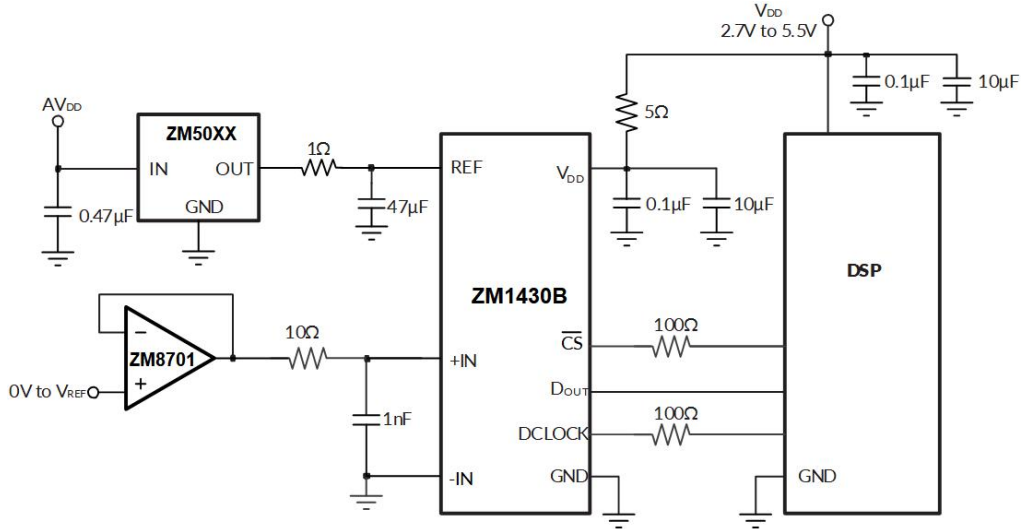


Figure 55. Basic Data Acquisition System: Example 1

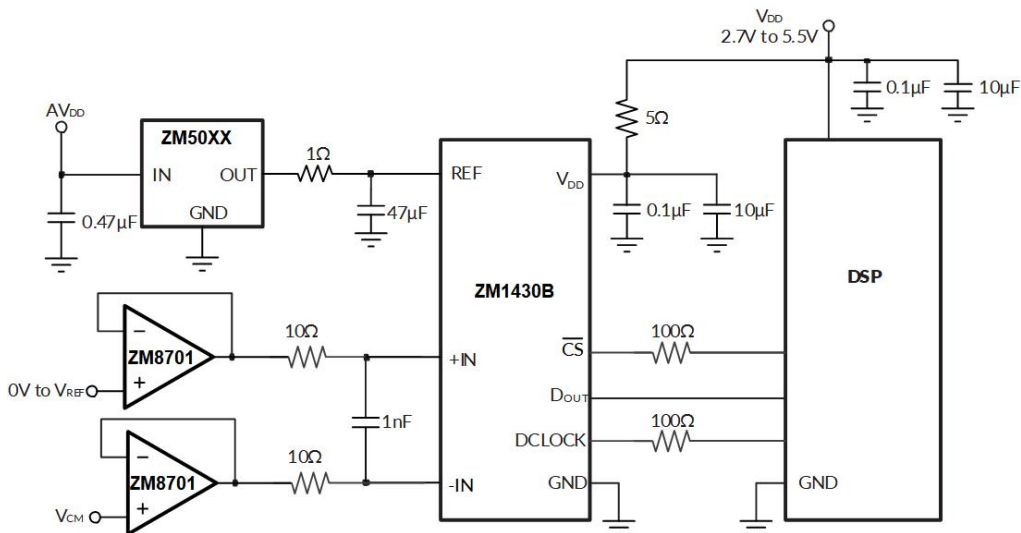
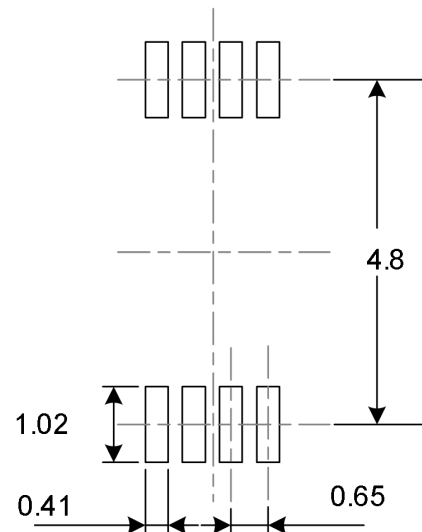
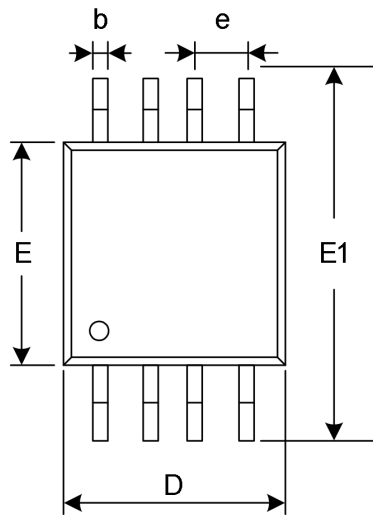


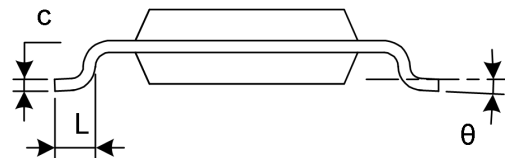
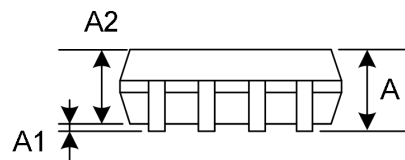
Figure 56. Basic Data Acquisition System: Example 2

12 Package Outline Dimensions

MSOP8⁽³⁾



RECOMMENDED LAND PATTERN (Unit: mm)



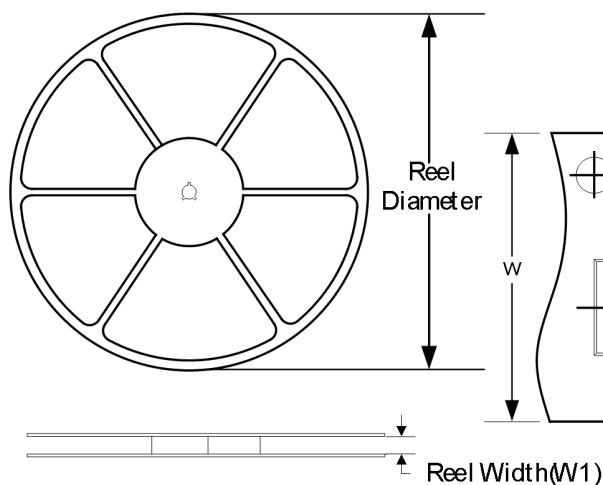
Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A ⁽¹⁾	0.820	1.100	0.032	0.043
A1	0.020	0.150	0.001	0.006
A2	0.750	0.950	0.030	0.037
b	0.250	0.380	0.010	0.015
c	0.090	0.230	0.004	0.009
D ⁽¹⁾	2.900	3.100	0.114	0.122
e	0.650(BSC) ⁽²⁾		0.026(BSC) ⁽²⁾	
E ⁽¹⁾	2.900	3.100	0.114	0.122
E1	4.750	5.050	0.187	0.199
L	0.400	0.800	0.016	0.031
θ	0°	6°	0°	6°

NOTE:

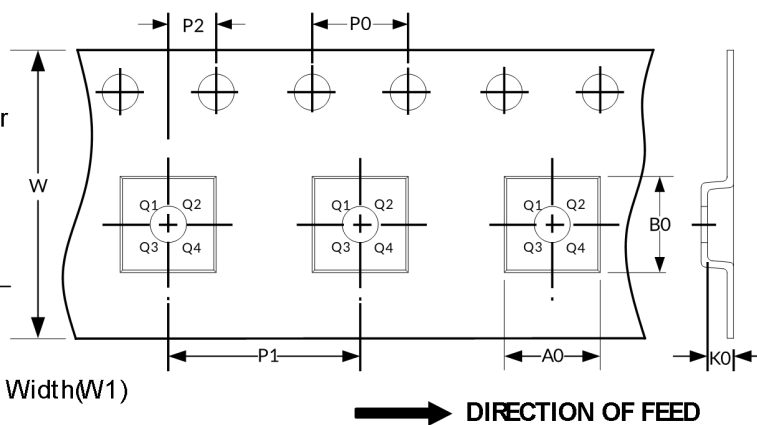
1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
3. This drawing is subject to change without notice.

13 Tape and Reel Information

REEL DIMENSIONS



TAPE DIMENSION



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width(mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
MSOP8	13"	12.4	5.20	3.30	1.50	4.0	8.0	2.0	12.0	Q1

NOTE:

1. All dimensions are nominal.
2. Plastic or metal protrusions of 0.15mm maximum per side are not included.

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